

The Impact Of Changing Magnetization Reactance Value On The Performance Of Three-Phase Induction Motor

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Abstract

The motor design for the three-phase induction motor under study was completed in this research using ANSYS Motorcad. The impact of variations in magnetization reactance (R_m) values on motor performance is often examined in these investigations. Efficiency, torque, current, and power factor were all taken into account in this work. Analysis and findings were achieved for magnetization reactance values between 0.5 and 2.5. Table 3.0 provides a summary of the findings, while Figures 1, 2, and 3 illustrate the effects. The flux distribution for the motor under study in the Motorcad environment is displayed in Figure 4.0. The findings show that at $R_m=0.5$, the torque is 3Nm, the efficiency is 61.5%, the current is 3.3A, and the power factor is 0.75%. Efficiency is 75%, torque drops to 2.55 Nm, current is 1.25A, and power factor is 0.8% at $R_m = 1.0$. Efficiency is 80%, torque is 1.82 Nm, current is 1.1 A, and power factor is 0.7% when $R_m=1.5$. 85% motor efficiency, 0.3 Nm torque, 1.0 A current, and a stable power factor are achieved with $R_m=2.5$. Since R_m values are components of magnetization current, they have an impact on no-load losses.

Keyword: Magnetization Reactance, Three-Phase Induction Motor, Motor Performance, Impedance Variation, Electrical Machine Analysis

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I. Introduction

The magnetizing resistance (R_m) is a critical parameter in induction motor design and operation, playing a significant role in determining the motor's performance characteristics. R_m represents the opposition to the magnetizing current in the motor, which is essential for creating the magnetic field that enables the motor to produce torque Okoro (2003), Eke, et al (2024). The magnetization reactance (X_m) is an important factor in the performance of a three-phase induction motor. It is a component of the motor's equivalent circuit that indicates the magnetizing current necessary to create a magnetic field in the stator and rotor of the motor. When the magnetization reactance changes, various aspects of the motor's performance are affected, including Perduková, et al (2020). The magnetization reactance controls how much magnetizing current is needed to generate a magnetic field in the motor. A greater X_m number indicates a larger magnetizing current, while a lower X_m value indicates a lower magnetizing current. The magnetizing current accounts for a sizable amount of the motor's no-load current draw. Liang et al (2017), Park, et al. (2024). The no-load current will grow as X_m increases. A drop in X_m , on the other hand, will result in a decrease in no-load current. The magnetizing current is roughly 90 degrees ahead of the voltage, resulting in a trailing power factor under no-load conditions. Changes in X_m can affect the magnitude of the reactive power component and, as a result, the power factor. The magnetization reactance affects the magnetic flux in the air gap of the motor. A larger X_m value causes more magnetic flux, which affects the motor's performance, torque, and efficiency. A significant magnetizing current is required to establish the magnetic field during motor starting. Changes in X_m can have an impact on the beginning current and torque. Park, et al (2024). In an induction motor, the magnetization reactance (X_m) is a rotor's equivalent circuit component. It denotes the reactance associated with the magnetizing current necessary to create a magnetic field in the rotor of the motor. The magnetization reactance influences no-load losses, which include iron losses in the rotor caused by the magnetizing current. When X_m grows, so do the iron losses, resulting in a decline in the motor's efficiency at full load. Reduced X_m , on the other hand, can boost efficiency by lowering these losses. Park, et al (2023). At low loads, core losses, including those produced by the magnetizing current, account for a larger proportion of total losses. The core losses grow as X_m increases, lowering the motor's efficiency at light loads. Under light load conditions, a lower X_m can aid in improving efficiency Aoyama, et al (2021). The magnetizing current travels through the rotor windings, causing copper losses. Higher magnetizing current and copper losses come from increasing X_m . This additional copper loss affects the efficiency of the motor. Reduced X_m , on the other hand, can reduce copper losses and improve overall efficiency. Banchhor, et al. (2018). The magnetizing current is required during motor startup to generate the initial magnetic flux in the rotor. A greater

X_m number raises the starting current, resulting in a poorer starting efficiency. Reducing X_m can help increase the motor's starting efficiency. Ejiofor, et al (2019), Dey et al (2008), Sengamalai, et al (2022). Because of the higher reactive component of the magnetizing current, a bigger X_m results in a lower power factor; a lower X_m value results in a better power factor, which raises overall efficiency reactance affects the motor's power factor. [Park, et al. (2024).

II. Mathematical Modelling And Equivalent Circuit Of Three-Phase Induction Motor

The dynamic model equations of the induction motor according to Okoro (2003) can be obtained from the dq0 equivalent circuit of the induction motor shown in Figure 1.1, figure 1.2.

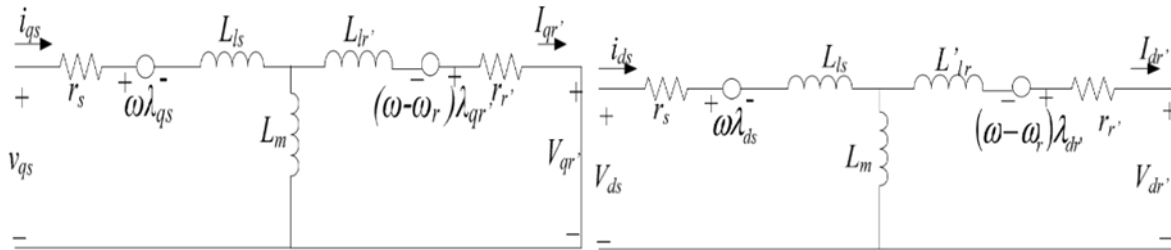


Fig. 1. D-axis equivalent circuit of an induction motor Okoro (2003)

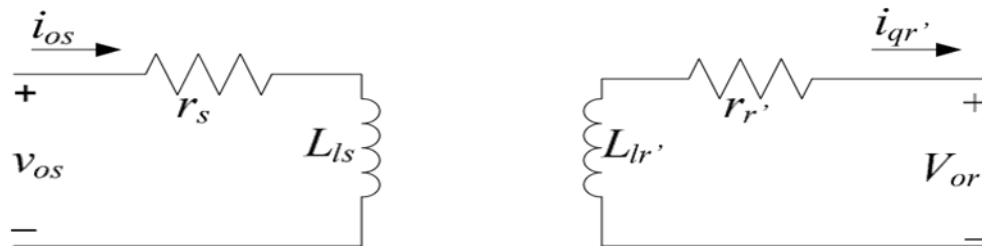


Fig. 1.2. Q-axis equivalent circuit of an induction motor Okoro (2003)

Under balanced conditions the three-phase stator voltage of an induction motor can be expressed as

$$V_a = \sqrt{2} V_{rms} \sin(\omega t) \quad (1)$$

$$V_b = \sqrt{2} V_{rms} \sin\left(\omega t - \frac{2\pi}{3}\right) \quad (2)$$

$$V_c = \sqrt{2} V_{rms} \sin\left(\omega t + \frac{2\pi}{3}\right) \quad (3)$$

These three-phase voltages are transformed into two-phase synchronously rotating reference frames of dq0 axis.

$$\begin{bmatrix} V_\alpha \\ V_\beta \end{bmatrix} = \frac{2}{3} \begin{bmatrix} 1 & \frac{1}{2} & -\frac{1}{2} \\ 0 & \frac{\sqrt{3}}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} V_a \\ V_b \\ V_c \end{bmatrix} \quad (4)$$

Then, the dq axis voltages

$$\begin{bmatrix} V_d \\ V_q \end{bmatrix} = \begin{bmatrix} \cos \theta & \sin \theta \\ -\sin \theta & \cos \theta \end{bmatrix} \begin{bmatrix} V_\alpha \\ V_\beta \end{bmatrix} \quad (5)$$

The instantaneous value of the stator and rotor currents of a three-phase induction motor is calculated by using the following matrix equations (6) and (7); Montaser et al (2025)

$$\begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix} = \begin{bmatrix} \cos \theta & -\sin \theta \\ \sin \theta & \cos \theta \end{bmatrix} \begin{bmatrix} i_d \\ i_q \end{bmatrix} \quad (6)$$

$$\begin{bmatrix} i_a \\ i_b \\ i_c \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \frac{1}{2} & -\frac{\sqrt{3}}{2} \\ -\frac{1}{2} & -\frac{\sqrt{3}}{2} \end{bmatrix} \begin{bmatrix} i_\alpha \\ i_\beta \end{bmatrix} \quad (7)$$

The stator and rotor voltage equation of the direct and quadrature (d-q) axis of an induction motor are given in equations (8), (9), (10), and (11). Eke, et al (2024)

$$V_{qs} = R_s i_{qs} + \frac{d}{dt} \lambda_{qs} + \omega_e \lambda_{qr} \quad (8)$$

$$V_{ds} = R_s i_{ds} + \frac{d}{dt} \lambda_{ds} - \omega_e \lambda_{qs} \quad (9)$$

$$V_{qr} = R_r i_{qr} + \frac{d}{dt} \lambda_{qr} + (\omega_e - \omega_r) \lambda_{qr} \quad (10)$$

$$V_{dr} = R_r i_{dr} + \frac{d}{dt} \lambda_{dr} - (\omega_e - \omega_r) \lambda_{dr} \quad (11)$$

Where, λ_{qr} , λ_{qs} , λ_{dr} , and λ_{ds} are the stator and rotor flux linkage of the d and q axis, these equations show the synchronously rotating reference frame for the two-phase d-q-axis of the induction motor. Okoro (2003) For squirrel cage induction motor, the rotor voltages V_{qr} , V_{dr} are set to zero, since the rotor cage bars are shorted. Therefore, the flux linkage equation can be expressed as equations (12), (13), (14), (15), (16), (17), and (18). Ali et al (2024), Montaser et al (2025)

$$\frac{d\lambda_{qs}}{dt} = \omega_b \left[V_{qs} - \frac{\omega_e}{\omega_b} \lambda_{ds} + \frac{R_s}{X_{ls}} (\lambda_{mq} - \lambda_{qs}) \right] \quad (12)$$

$$\frac{d\lambda_{ds}}{dt} = \omega_b \left[V_{ds} + \frac{\omega_e}{\omega_b} \lambda_{qs} + \frac{R_s}{X_{ls}} (\lambda_{md} - \lambda_{ds}) \right] \quad (13)$$

$$\frac{d\lambda_{qr}}{dt} = \omega_b \left[V_{qs} - \left(\frac{\omega_e - \omega_r}{\omega_b} \right) \lambda_{qr} + \frac{R_r}{X_{lr}} (\lambda_{mq} - \lambda_{qr}) \right] \quad (14)$$

$$\frac{d\lambda_{dr}}{dt} = \omega_b \left[V_{ds} + \left(\frac{\omega_e - \omega_r}{\omega_b} \right) \lambda_{dr} + \frac{R_r}{X_{lr}} (\lambda_{md} - \lambda_{dr}) \right] \quad (15)$$

Where;

$$\lambda_{mq} = X_{ml} \left[\frac{\lambda_{qs}}{X_{ls}} + \frac{\lambda_{qr}}{X_{lr}} \right] \quad (16)$$

$$\lambda_{md} = X_{ml} \left[\frac{\lambda_{ds}}{X_{ls}} + \frac{\lambda_{dr}}{X_{lr}} \right] \quad (17)$$

$$X_{ml} = \frac{1}{\left(\frac{1}{X_m} + \frac{1}{X_{ls}} + \frac{1}{X_{lr}} \right)} \quad (18)$$

To obtain the currents, therefore substitute the value of flux linkages as shown in equations (19), (20), (21), and (22).

$$i_{qs} = \frac{1}{X_{ls}} (\lambda_{qs} - \lambda_{mq}) \quad (19)$$

$$i_{ds} = \frac{1}{X_{ls}} (\lambda_{ds} - \lambda_{md}) \quad (20)$$

$$i_{qr} = \frac{1}{X_{lr}} (\lambda_{qr} - \lambda_{mq}) \quad (21)$$

$$i_{dr} = \frac{1}{X_{lr}} (\lambda_{dr} - \lambda_{md}) \quad (22)$$

Based on the above equations, the torque and rotor speed can be determined as equations (25) and (26). Ferreira, et al. (2008).

$$T_e = \frac{3}{2} \left(\frac{p}{2} \right) (\lambda_{qr} i_{dr} - \lambda_{dr} i_{qr}) \quad (25)$$

$$\omega_e = \int \frac{p}{2} (T_e - T_L) \quad (26)$$

III. Methods Of Calculating Magnetsing Reactance

The magnetizing reactance (X_m) of an induction motor can be calculated using various methods. Here are some common methods

i. Using Magnetizing Inductance Park, et al (2024).

$$X_m = \omega * L_m \quad (27)$$

where: X_m = magnetizing reactance, ω = angular frequency ($2 * \pi * f$), L_m = magnetizing inductance

ii. Using Motor Parameters

$$X_m = (V_1 / I_m) * \sin(\theta_m) \quad (28)$$

where: V_1 = stator voltage, I_m = magnetizing current, θ_m = magnetizing current angle

iii. Using Equivalent Circuit Parameters

$$X_m = X_{lm} * (1 + (R_2 / X_2)) \quad (29)$$

where: X_{lm} = stator magnetizing reactance, R_2 = rotor resistance, X_2 = rotor reactance

iv. Using Design Data

$$X_m = (4 * \pi * f * N^2 * A) / (g * l) \quad (30)$$

where: f = frequency, N = number of turns, A = cross-sectional area of the magnetic circuit, g = air gap length
 l = length of the magnetic circuit

v. Using No-Load Test Data

$$X_m = V_1 / (I_0 * \sin(\theta_0)) \quad (31)$$

where: V_1 = stator voltage, I_0 = no-load current, θ_0 = no-load current angle

These equations illustrate the different methods for calculating X_m , which is an important parameter in induction motor design and analysis. The choice of method depends on the available data and the specific application.

IV. Experimental Setup

Ammeters, voltmeters, and wattmeters were attached to measure the current, voltage, and power from the 1.5KW three-phase induction motor, which was powered by a three-phase variable voltage source. For the no-load test, the motor was let to operate.

Table 1.0: Specifications of the studied motor

Part	Parameter	Value
Rated parameters	Rated voltage	380 V
	Rated current	3.80 A
	Rated speed	1500 rpm
	Power output	1.5 kW
	slot numbers	34
Stator	Outer diameter	120 mm
	Inner diameter	75 mm
	Pole embrace	0.45
	Yoke thickness	12 mm
	Frame	Mec100
	Pole number	4
	Rotor bar	28
Rotor	Pole embrace	0.3
	Yoke thickness	9 mm
	Inner diameter	30mm

Table 1.0 shows the rated motor specifications; these values were used in design of the motor in motor experiments and simulations were done and results gotten.

Table 2.0: Results From No Load Test

Voltage(V)	Current (Ia)	Current (Ib)	Current (Ic)	Average current	Input Power (P)	Speed (N) RPM
50	0.58	0.61	0.56	0.58	38.6	1452
70	0.49	0.55	0.42	0.48	44.8	1513
90	0.43	0.48	0.46	0.45	54.0	1525
110	0.43	0.61	0.49	0.51	74.9	1480
130	0.48	0.75	0.59	0.60	104.0	1488
150	0.60	0.71	0.71	0.67	134.0	1491
170	0.73	0.79	0.79	0.74	167.7	1493
190	0.79	0.90	0.90	0.86	217.9	1494
210	0.93	0.95	0.95	0.94	263.2	1505
230	0.97	1.06	1.06	1.03	315.9	1465
250	0.95	1.27	1.27	1.16	386.7	1464
270	1.11	1.40	1.40	1.39	500.5	1497
290	1.21	1.59	1.59	1.46	561.8	1497
310	1.24	1.84	1.84	1.64	678.0	1498
330	1.40	1.77	1.77	1.65	726.2	1498
350	1.54	1.97	1.97	1.82	849.5	1498
370	1.58	2.05	2.05	1.89	932.6	1498
380	1.72	2.18	2.18	2.02	1027	1498
390	2.02	2.24	2.24			1498

V. Result Presentation And Discussion $R_m=0.5$

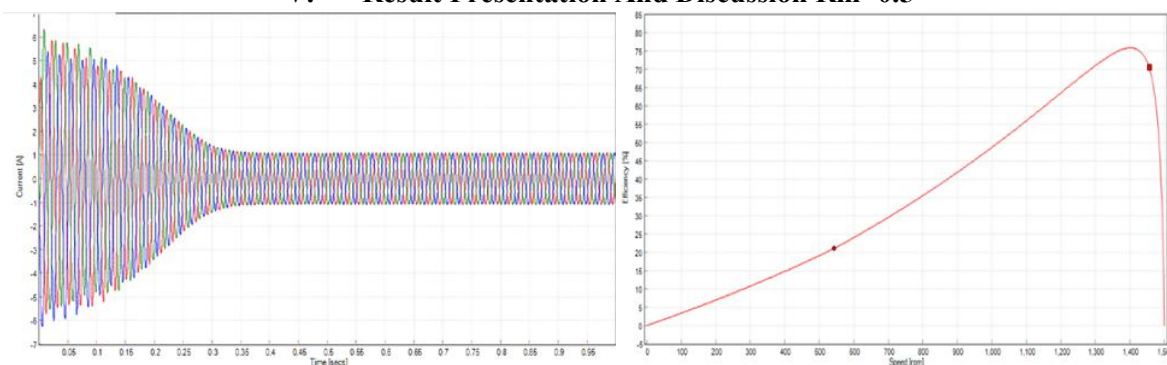


Fig. 1.0: graph efficiency against speed at $R_m=1$

Fig.1.2: graph of Current against Time when $R_m=1$.

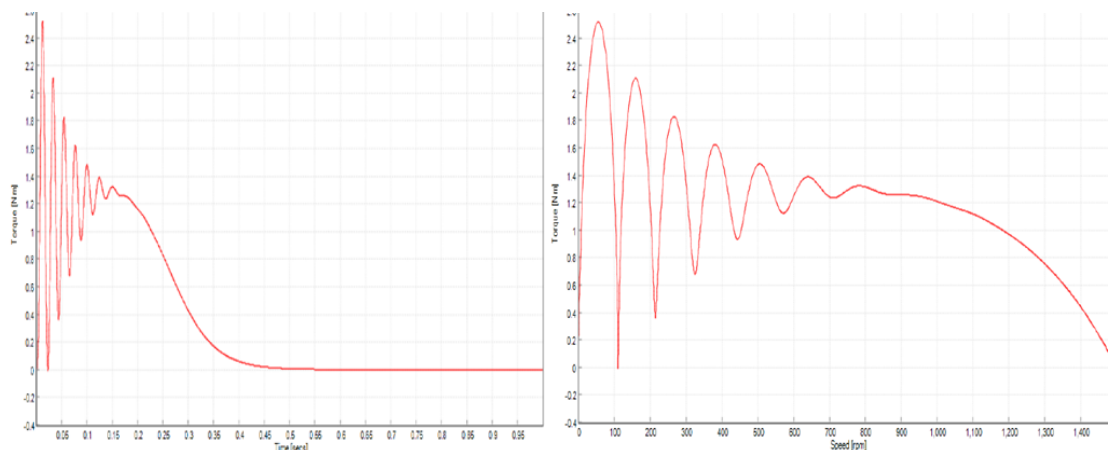


Fig. 1.3: graph of torque against time when $R_m=1$.

Fig. 1.4: graph of torque against speed when $R_m=1$.

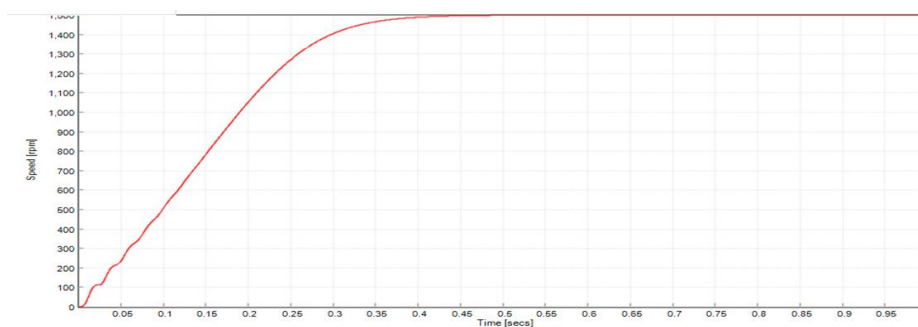


Fig. 1.5: Graph of speed against time

AT $R_m = 1.5$

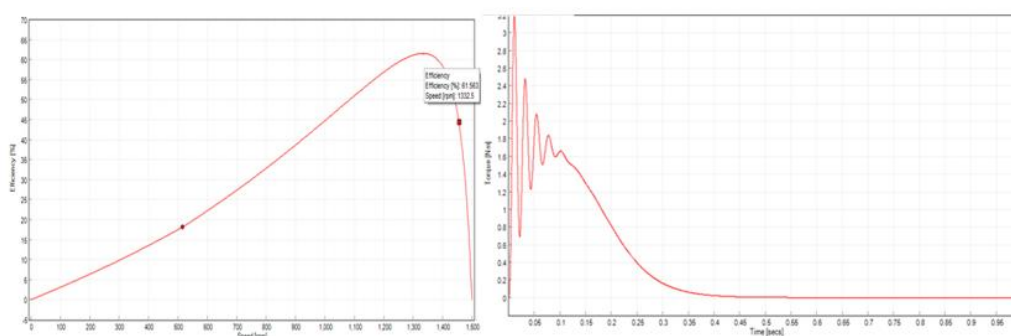


Fig. 2.0: graph efficiency against speed at $R_m=0.5$

Fig. 2.2: graph of torque against time when $R_m=0.5$.

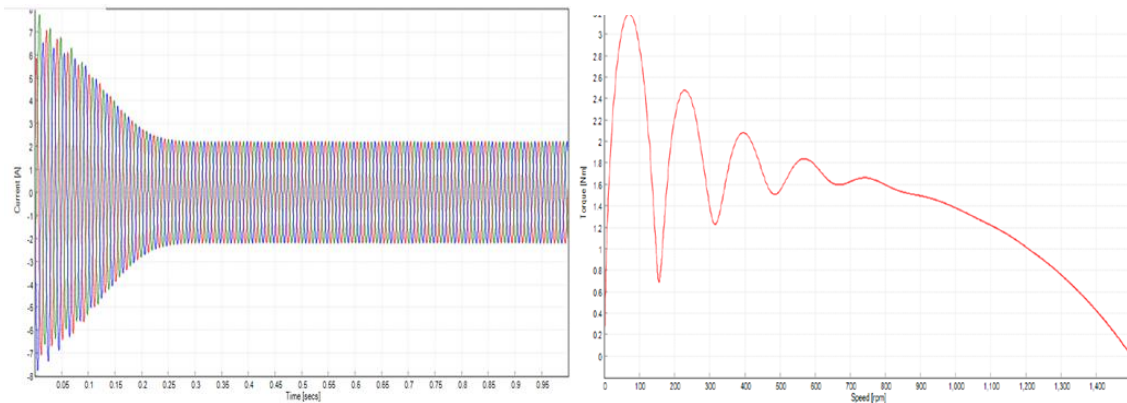


Fig. 2.3: graph of Current against Time when $R_m=0.5$.

Fig. 2.4: graph of torque against speed when $R_m=0.5$.

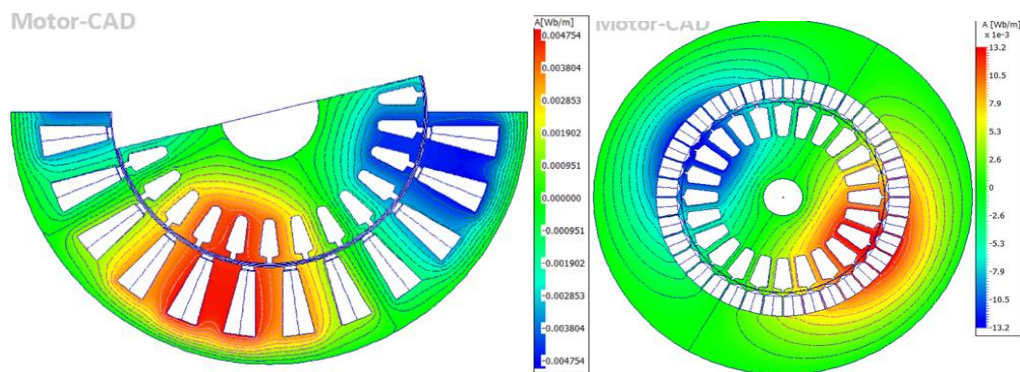


Figure 4.1: E-magnetic model showing flux distribution

Fig. 4.2: E-magnetic model showing flux distribution

AT RM = 2.5

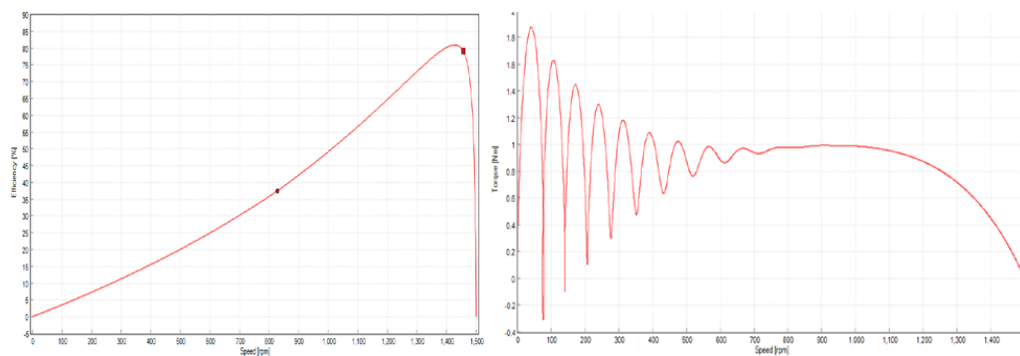


Fig. 3.1: Graph efficiency against speed at $R_m=1.5$

Fig. 3.2: graph of torque against speed when $R_m=1.5$.

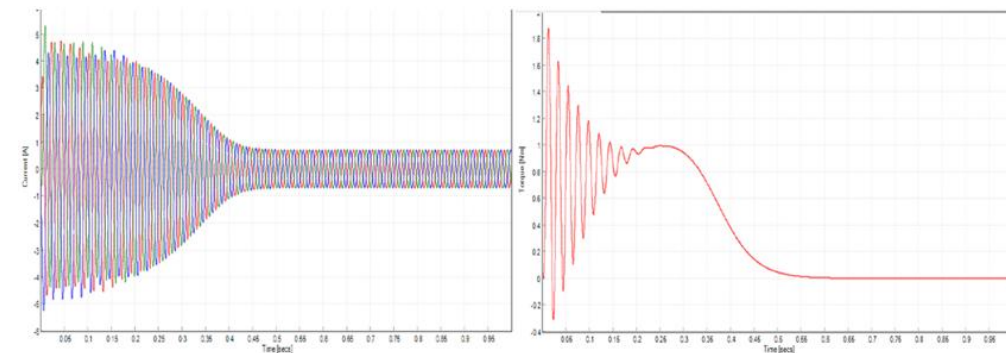


Fig. 3.3: graph of Current against Time when $R_m=1.5$.

Fig. 3.4: graph of torque against time when $R_m=1.5$.

Table 2.0: motor performance result summary at the rated speed of 1500rpm from simulation

Magnetization reactance (Rm) values	0.5	1.0	1.5	2.5
Efficiency (%)	61.5	75	80	85
Torque (N/m)	3.2	2.55	1.82	0.3
Current (A)	3.3	1.2	1.1	1.0
Power factor (%)	75	80	70	76

VI. Discussion Of Results

The motor rating values utilized in the studies are displayed in Table 1. This covers the parameters of the rotor and stator. Following the process, no load test was performed. Table 2.0 displays the experiment's findings. The voltage (V), current (A), power (W), and motor speed (rpm) at each voltage value are changed in this table. The rated voltage and speed are 2.22A and 1200w, respectively. After a 20-minute break, the trial findings were collected. MotorCAD software was used to further analyze the effects of changing Rm values on motor performance using the rating parameters listed in table 1.0. The effect's outcomes are displayed in Figures 1, 2, and 3. Rm can have values between 0.5, 1.0, 1.5, and 2.5. Figures 1.1, 1.2, 1.3, 1.4, and 1.5 consist of the efficiency vs speed, current versus time, torque versus speed, torque versus time, and speed versus time graphs, respectively, presented in Figure 1. Figure 1.1 shows that at a speed of 1450 rpm, the efficiency is 75%. The current in Fig. 1.2 starts at 5.6A at 0.01 seconds and stabilizes at 1.8A at 0.3 seconds. At 0.5 seconds and 75 rpm, the beginning torque in Figures 1.3 and 1.4 is 2.55 Nm. Figure 2.1 shows an efficiency of 61.56% at 1332 rpm when Rm is 0.5. Figure 2.2 shows a torque of 3.2 Nm at startup, Figure 2.3 shows a current of 7.8A at 0.01 seconds and 2.8A at 2 seconds. The efficiency in Fig. 3.1 is 80%. Rm was 1.5 when this took place. at this time, the torque is 1.82Nm, the beginning current is 5.2A, and at 4.5seconds the current is 0.8A

VII. Conclusion

The magnetization reactance (Xm) is critical in calculating the core and copper losses in a three-phase induction motor. Higher Xm values result in greater core and copper losses and lower efficiency, particularly during light and full load situations. Reducing Xm, on the other hand, can help enhance the motor's efficiency performance by lowering losses and increasing the power factor. On the other hand, the best value of Xm is the result of a trade-off between different elements, and motor design requires careful consideration of multiple parameters to obtain the desired efficiency and performance for specific applications. A three-phase induction motor's magnetization reactance (Xm) significantly affects its torque performance. The magnetization reactance is a part of the rotor's equivalent circuit and represents the reactance associated with the magnetizing current required to establish the magnetic field in the motor's rotor. Here's how changes in the magnetization reactance (Xm) impact the torque performance of a three-phase induction motor are as follows; During motor starting, a significant amount of torque is required to overcome inertia and accelerate the load. The starting torque is directly related to the magnetizing current and the magnetization reactance (Xm). Higher Xm values lead to higher magnetizing current, resulting in a stronger magnetic field and increased starting torque. Conversely, reducing Xm can reduce the magnetizing current and, consequently, the starting torque. The pull-up torque is the maximum torque a motor can produce during starting before it slips out of synchronism. It is essential for starting loads with high inertia. A higher Xm value generally leads to higher pull-up torque, while a lower Xm reduces the pull-up torque. The breakdown torque is the maximum torque an induction motor can produce without stalling or tripping the overload protection. Changes in Xm can affect the breakdown torque, but the impact is not as straightforward as with starting torque. Other design parameters, such as rotor resistance and leakage reactance, also play significant roles in determining the breakdown torque. The magnetization reactance influences the magnetic field distribution in the motor. A higher Xm can reduce torque ripple, resulting in smoother operation and reduced mechanical vibrations. The synchronous speed of an induction motor is given by the formula: Synchronous Speed (RPM) = (120 * Frequency) / Number of Poles. Changes in Xm can slightly affect the effective number of poles, thereby impacting the synchronous speed of the motor. The magnetization reactance affects the magnetizing current and, consequently, the overall motor performance at full load. Higher Xm values can lead to higher copper losses due to increased magnetizing current, affecting the motor's full-load torque. It's essential to optimize the magnetization reactance (Xm) along with other design parameters to achieve the desired torque performance for specific applications. A well-balanced motor design involves considering torque requirements, efficiency, starting characteristics, and other factors to ensure optimal performance under various operating conditions. Manufacturers use advanced design tools and simulations to achieve a given application's most suitable magnetization reactance and overall motor performance

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