

VLSI Design of Secure Cryptographic Algorithm

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ABSTRACT : Light weight cryptography (LWC) is an emerging research area which has to deal with the trade-off among security, cost, and performance. In this paper we present the idea and list some types of LWC algorithms. Hummingbird is a novel ultra lightweight cryptographic algorithm targeted for resource constrained devices like RFID tags, smart cards and wireless sensor nodes. The hybrid model of Hummingbird is explained keeping the constraint devices in mind and thus resulting in an easier software implementation. The paper presents the algorithms for the encryption as well as decryption process and shows some simulation results performed on Xilinx.

Keywords: Lightweight cryptography, resource constrained devices, Hummingbird, RFID

I. INTRODUCTION

RFID (Radio Frequency Identification) tags have numerous applications, ranging from supply chain optimization to ehealth. Basically, RFID tags consist of a transponder and an antenna and are able to remotely send and receive data from an RFID host or reader device. In general, RFID tags can be divided into passive and active devices: active tags provide their own power supply (i.e. in form of a battery), whereas passive tags solely rely on the energy of the carrier signal transmitted by the reader device. Passive RFID devices are not only much less expensive, but also require less chip size and have a longer life cycle [1].

Our proposed DESL algorithm and its low-power, size optimized implementation aims at very constrained devices such as passive RFID tags. Providing cryptographic primitives (esp. encryption) at extremely low cost is of paramount importance for securing RFID applications. Thus far, there have been two approaches for providing cryptographic primitives for such situations:

Optimized low-cost implementations for standardized and trusted algorithms, which means in practice in essence block ciphers such as AES [2], see e.g., [3].

Design new ciphers with the goal of having low hardware implementation costs²

The best known light-weight AES implementation [3] requires 3400 gates and draws a maximum current of 3.0 μ A @ 100 kHz. This AES design is based on a byte-per-byte serialization, which only requires the implementation of a single S-box [4] and achieves one encryption within 1032 clock cycles (= 10.32ms @ 100 kHz). Unfortunately, the ISO/IEC 18000 standard requires that the latency of a response of an RFID tag does not exceed 320 μ s, which is why [3] proposes a slightly modified challenge-response protocol based on interleaving. The problem with this approach is that AES like most modern block ciphers was primarily designed with good software implementation properties in mind, and not necessarily with hardware-friendly properties. The only known cipher that was designed with a strong focus on low hardware costs is the Data Encryption Standard, DES [5]. If we compare a standard, one-round implementation of AES and DES, the latter consumes about 6% (!) of the logic resources of AES, while having a shorter critical path [6], [7]. Therefore, we decided to follow the second approach and modify the hardware efficient and very well investigated cipher DES.

The main design ideas of the new cipher family, which are either original DES efficiently implemented or a variant of DES, are:

- 1) Use of a serial hardware architecture which reduces the gate complexity.
- 2) Optionally apply key-whitening in order to render brute force attacks impossible.
- 3) Optionally replace the 8 original S-Boxes by a single one which further reduces the gate complexity.

If we make use of the first idea, we obtain a lightweight implementation of the original DES algorithm which consumes about 35% less gates than the best known AES implementation [3]. To our knowledge, this is the smallest reported DES implementation, trading area for throughput. The implementation requires also about 86% fewer clock cycles for encrypting of one block than the serialized AES implementation in [3] (1032 cycles vs. 144) which makes it easier to use in standardized RFID protocols. However, the security provided is limited by the 56 bit key. Brute forcing this key space takes a few months and hundreds of PCs in software, and only a

few days with a special-purpose machine such as COPACOBANA [8]. Hence, this implementation is only relevant for application where short-term security is needed, or where the values protected are relatively low. However, we can imagine that in certain low cost applications such a security level is adequate.

In situation where a higher security level is needed key whitening, which we define here as follows: $DESX_{k_1, k_2}(x) = K_2 \oplus DES_{k_1}(x)$, can be added to standard DES, yielding DESX. The bank of XOR gates increases the gate count by about 14%. The best known key search attack uses a time-memory trade-off and requires 2120 time steps and 264 memory locations, which renders this attack entirely out of reach. The best known mathematical attack is linear cryptanalysis [9]. LC requires about 243 chosen Cipher text blocks together with the corresponding plaintexts. At a clock speed of 500 kHz, our DESX implementation will take more than 80 years, so that analytical attacks do not pose a realistic threat. Please note that parallelization is only an option if devices with identical keys are available.

In situations where extremely light-weight cryptography is needed, we can further decrease the gate complexity of DES by replacing the eight original S-Boxes by a single new one. This light-weight variant of DES is named DESL and has a brute-force resistance of 256. In order to strengthen the cipher, key whitening can be applied yielding the ciphers DESXL. The crucial question is what the strength of DESL and DESXL is with respect to analytical attacks. We are fully aware that any changes to a cipher might open the door to new attacks, even if the changes have been done very carefully and checked against known attacks. Hence, we believe that DESL (or DESXL) should primarily not be viewed as competitors to AES, but should be used in applications where established algorithms are too costly. In such applications which have to trade security (really: trust in an algorithm) for cost, we argue that it is a cryptographically sounder approach to modestly modify a well studied cipher (in fact, the world's best studied crypto algorithm), rather than designing a new algorithm altogether.

Light Weight Implementation of DES and DESL

In this section we present a size-optimized design of DES, which is smaller than any previous implementations of DES to our knowledge.

Design Considerations and Implementation of DES

The overall architecture of our size-optimized DES implementation is depicted in Figure 1. Our design basically consists of five core modules: mem left, mem right, key schedule, controller, and sbx. The controller module manages all control signals in the ASIC based on the finite state machine depicted in Figure 3. In the key schedule module all DES round keys are generated. It is composed of a 56-bit register, an input multiplexor, and an output multiplexor to select the right fraction of the round key. The mem left module consists of eight 4-bit wide registers, each composed of D-flip-flops. The mem right module is similar to the mem left module with slight differences. It also consists of eight 4-bit wide registers, but it has different input and output signals: instead of a 4-bit wide output it has a 6-bit wide output, due to the expansion function of DES. The sbx module consists of eight S-boxes of the DES algorithm and an output multiplexor. The S-boxes are realized in combinatorial logic, i.e. a sum of products (SOP) [10]. Furthermore, we investigated whether there exists a correlation between the design criteria (i.e. linear and differential characteristics) and the required number of logic gates (AND, OR, NOT) for each S-box. However, we did not observe a significant deviation for any S-box.

Figure 1 shows the data path of our serialized DES design. The 56-bit key is stored in the key flip-flop register after the PC1 and LS1 permutations have been applied. The plaintext is first confused using the Initial Permutation (IP), then, it is split into two 32-bit inputs for the modules mem left and mem right, respectively. The input of mem left is modified by the inverse of the P permutation and stored in the registers of the modules mem left and mem right in one cycle. Next, the output of the last register in mem right is both stored in the first register of mem right and expanded to six bits. After an XOR operation with the appropriate block of the current round key, this expanded value is processed by the sbx module, which is selected by the count signal, provided by the controller module. Finally, the result is XORed with the output of the mem left module, and stored in the first flip-flop of the mem left module. This is repeated eight times, until all 32 bit of the right half are processed.

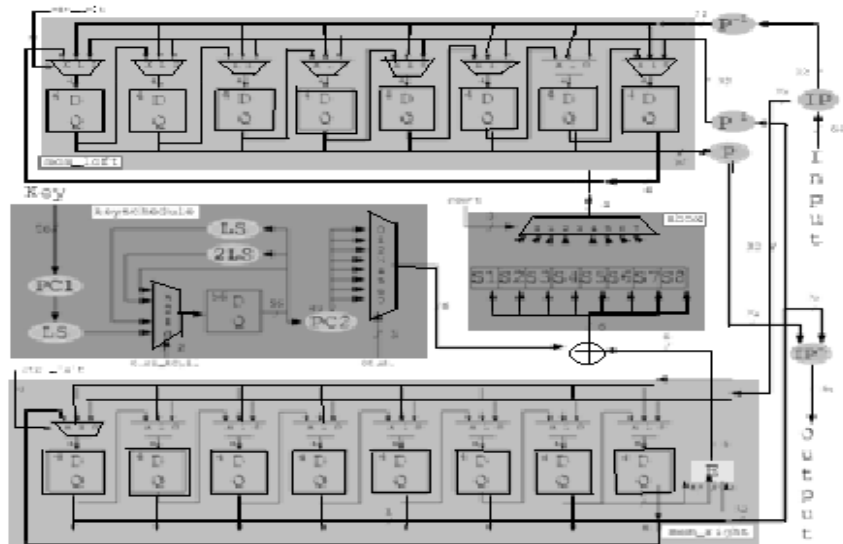


Fig.1. Data path of the serialized DES ASIC

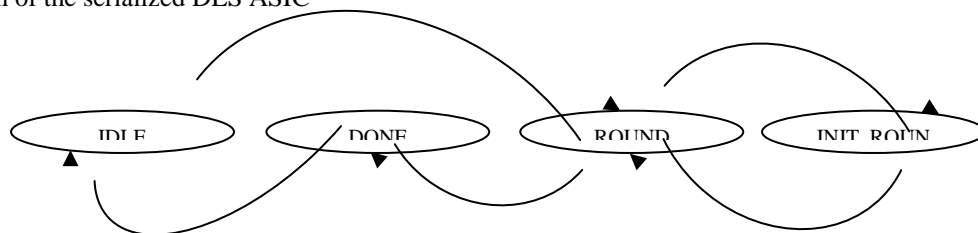


Fig. 2. Finite State Machine of the DES ASIC

In our design, we applied the P permutation in each ninth clock cycle. Because the P-1 permutation is applied before the left 32-bit half L_i is stored in the mem left module, we perform the (P) permutation on the resulting right half R_{i+1} : $R_{i+1} = P(P-1(L_i) \oplus S(E(R_i \oplus K_i)))$, where L_i denotes the left half, R_i denotes the right half, and K_i denotes the round key.

By reducing the data path from a 32-bit bus to a 4-bit bus, only $6 \times 10 + 4 \times 10 = 100$ transistors (25 GE) are needed for the XOR operations, compared to $48 \times 10 + 32 \times 10 = 800$ (200 GE) transistors in a not-serialized design. This saving comes with the disadvantage of two additional multiplexors, each one for the round key (288 transistors, 72 GE) and for the S-box output (192 transistors, 48 GE). As we will see in Section II-C, our DESL algorithm does not need an output multiplexor in the sbox module. Once all eight 4-bit blocks of both halves have been processed, they are concatenated to two 32-bit wide outputs of the modules mem left and mem right. The output of the module mem left is transformed by the P permutation and stored as the new content of the mem right module, while the output of the mem right module is stored as the new content of the mem left module.

This execution flow repeats another 15 rounds. Finally, both outputs of the memory modules mem left and mem right are concatenated to a 64-bit wide output. This output is confused by the Inverse Initial Permutation (IP-1), which results in a valid cipher text of the DES algorithm.

We used Synopsys Design Vision V-2004.06-SP2 to map our DES design to the Artisan UMC 0.18 μ m L180 Process 1.8-Volt Sage-X Standard Cell Library and Cadence Silicon Ensemble 5.4 for the Placement & Routing-step. It takes 144 clock cycles to encrypt one 64-bit block of plaintext. For one encryption at 100 kHz the average current consumption is 1.19 μ A and the throughput reaches 5.55 KB/s.

II. FURTHER OPTIMIZATION CONSIDERATIONS

In our DES ASIC design registers take up the main part of chip size (33.78%), followed by the S-boxes (32.11%), and multiplexors (31.19%). Chip size of registers and multiplexors can not be minimized any further, hence we thought about further possibilities to optimize the chip size of the S-boxes.

While it does not seem to be possible to find better logic minimizations of the original DES S-boxes, there have been other approaches to alter the S-boxes, e.g. key-dependent S boxes [11], [12] or the so-called si DES [13]–[15]. While all these approaches, despite the fact that some of them have worse cryptographic properties than DES [16], just change the content and not the number of S-boxes. To the best of our knowledge, no DES variant has been proposed in the past which uses a single S-box, repeated eight times.

Our work regarding a strengthened S-box is based on the original design criteria for DES as published by Coppersmith [17] and the work of Kim et al. [13]–[15] where several criteria for DES type S-boxes are presented to strengthen the resistance against differential and linear cryptanalysis, and the Davies-Murphy-Attack [18].

| S | | | | | | | | | | | | | | | |
|----|---|----|----|----|---|----|----|----|----|----|---|----|----|----|----|
| 14 | 5 | 7 | 2 | 11 | 8 | 1 | 15 | 0 | 10 | 9 | 4 | 6 | 13 | 12 | 3 |
| 5 | 0 | 8 | 15 | 14 | 3 | 2 | 12 | 11 | 7 | 6 | 9 | 13 | 4 | 1 | 10 |
| 4 | 9 | 2 | 14 | 8 | 7 | 13 | 0 | 10 | 12 | 15 | 1 | 5 | 11 | 3 | 6 |
| 9 | 6 | 15 | 5 | 3 | 8 | 4 | 11 | 7 | 1 | 12 | 2 | 0 | 14 | 10 | 13 |

Table 1 Improved DESL S-box

The mathematical background of this part would go beyond the scope of this special session. Therefore we briefly list our conditions below:

If two inputs to an S-box differ in their first bit and are identical in their last two bits, the two outputs must not be the same.

$$|S_b^W(a)| \leq 28 \text{ for all } a \in GF(2)^6, b \in GF(2)^4.$$

$$S_b^W(a) \leq 4 \text{ for all } a \in GF(2)^6, b \in GF(2)^4 \text{ with } wt(a) = wt(b) = 1.$$

$$S_b^W(a) \leq 16 \text{ for all } a \in GF(2)^6, b \in GF(2)^4 \text{ with } wt(a), wt(b) \leq 2.$$

$$|S_{b_1}^W(a)S_{b_2}^W(a)| \leq 240 \text{ for all } a \in GF(2)^6, b_1, b_2 \in GF(2)^4 \text{ with } wt(a) = 1, wt(b_1 + b_2) = 1.$$

$$S_b^W(a) = 0 \text{ for all } a \in \{(010000), (000010)\}, b \in GF(2)^4 \text{ with } wt(b) = 1$$

$$|S_{b_1}^W(000010)S_{b_2}^W(000010)| = 0 \text{ for all } b_1, b_2 \in GF(2)^4 \text{ with } wt(b_1 + b_2) = 1$$

$$S_{(0100)}^W(000100) = 0$$

$S_b^W(a)$ denotes the Walsh-coefficient and $wt(x)$ denotes the Hamming weight of x . We randomly generated S-boxes, which fulfill the original DES criteria (S-1), (S-3), (S-4), (S-5), (S-7) (see [17]), and the above listed conditions 1 to 8. It can be shown, that S-boxes that fulfill these criteria are immune to linear and differential cryptanalysis, and the Davies-Murphy-Attack. Table I shows the S-box, which is used in DESL.

III. IMPLEMENTATION OF DESL

The main difference between DESL and DES lies in the f -function. We substituted the eight original DES S-boxes by a single but cryptographically stronger S-box, which is repeated eight times. Furthermore, we omitted the initial permutation (IP) and its inverse (IP⁻¹), because they do not provide additional cryptographic strength, but at the same time require area for wiring. The design of our DESL algorithm is exactly the same as for the DES algorithm, except for the (IP) and (IP⁻¹) wiring and the sbox module. The changed sbox module implements only one S-box. As one can see in Figure 2, this module neither needs the count control signal nor an output multiplexor, which saves another 192 transistors (48GE). It takes 144 clock cycles to encrypt one 64-bit block of plaintext. For one encryption at 100 kHz the average current consumption is 0.89 μ A and the throughput reaches 5.55 KB/s. All results are summarized in Table II.

| | Gate total | Equiv. rel. | Cycles/block | μ A at 100 kHz | Process μ m |
|---------------|------------|-------------|--------------|--------------------|-----------------|
| DESL | 1848 | 1 | 144 | 0.89 | 0.18 |
| DES | 2309 | 1.25 | 144 | 1.19 | 0.18 |
| DESX | 2629 | 1.42 | 144 | - | 0.18 |
| DESXL | 2168 | 1.17 | 144 | - | 0.18 |
| AES-128[3] | 3400 | 1.84 | 1032 | 3.0 | 0.35 |
| TRIVIUM [20] | 2599 | 1.41 | - | - | 0.13 |
| GRAIN-80 [20] | 1294 | 0.70 | - | - | 0.13 |
| HIGHT [21] | 3048 | 1.65 | 34 | - | 0.25 |

Table II Comparison of efficient ciphers based on gate count, clock cycle, and current consumption

IV. RESULTS AND CONCLUSIONS

We presented our implementation results of DES and DESL in Section II. Table II shows that our DESL cipher needs 20% less transistors compared with our DES implementation and 38% less transistors compared with the implementation in [6]. This table also shows, that DESL uses 25% less average current than DES. In comparison with the AES design in [3], our design needs 45% less gate equivalents and 86% less clock cycles as shown in Table II. Note that the AES design in [3] was implemented in a 0.35 μ m standard cell technology, whereas our design was implemented in a 0.18 μ m standard cell technology. Therefore a fair comparison is only possible with regard to the gate equivalences. Regarding area consumption, our DESL is competitive even to stream ciphers recently proposed within the eSTREAM project [19]. More interesting, DESL would be the second smallest stream cipher in terms of gate count compared to all eSTREAM candidates (see Table II). Due to the low current consumption and the small chip size required for our DESL design, it is especially suited for resource limited applications, for example RFID tags and wireless sensor nodes.

Finally, we can conclude, that DESL is more secure against linear and differential cryptanalysis and the Davies-Murphy attack, more size-optimized, and more power efficient than DES, which makes it especially suited for RFID applications. Furthermore, DESL is worth to be considered as an alternative for stream ciphers.

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